

Lecture 4

Converters II - Overlap

E3002 Power Electronics

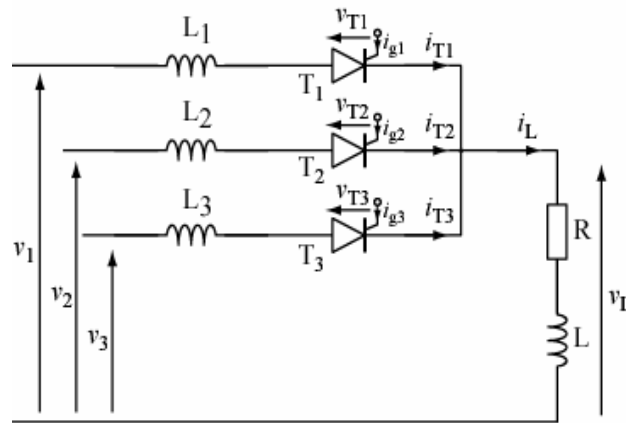
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Objectives

- Finite inductance in AC supply lines causes overlap in which case transfer of load current from one device to another in converter does not take place instantaneously
- We determine expressions for duration of overlap and calculate effect of overlap on mean load voltage of converter
- We model effect of overlap on converter output voltage by simple circuit consisting of ideal voltage source and resistor, whose value depends only on AC supply inductance and supply frequency

Introduction

- 3-phase, half-wave bridge with source inductance in AC supply lines
 - Consider instant of firing thyristor T_2 with T_1 conducting:



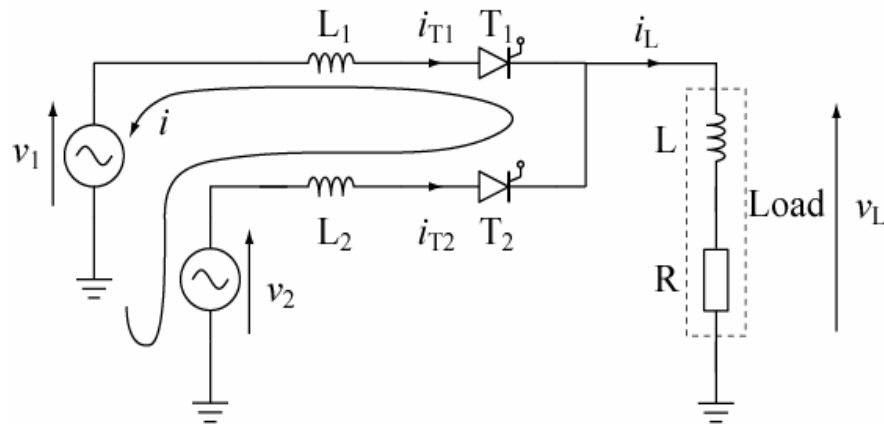
- When T_2 is fired, current I_{T1} can not fall immediately to zero and I_{T2} can not increase to equal I_L immediately.
- This is because voltage and current of inductor are related according to:

$$V_L = L \frac{dI_L}{dt}$$

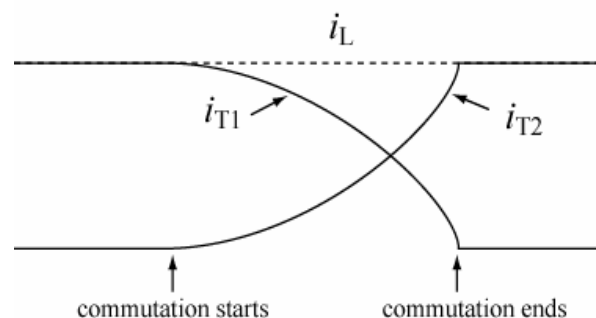
- Infinite dI_L/dt requires infinite voltage: fastest dI_L/dt limited by supply voltage
- Load current switches from T_1 to T_2 in finite time period with both T_1 and T_2 are conducting
- Overlap affects load voltage waveform and hence mean load voltage
- In practice, supply inductance < 1 mH; sufficient to have significant effect

Definition of overlap parameters

- Consider transfer of load current from T_1 to T_2
- AC supply V_3 and thyristor T_3 are not involved in this transfer of current
- Show AC supplies V_1 to V_2 with their inductance L , T_1 , T_2 and load:



- We assume a constant load current I_L
- Transfer of load current from T_1 to T_2 during switching:



- Actual dynamic currents = initial conditions ($I_{LT1} = I_L$, $I_{T2} = 0$) + current i circulates between T_1 and T_2 and AC supply sources

- Circulating current i increases from zero at instant of firing T_2 until it equals load current I_L ; at this point current I_{T1} has become zero when this thyristor is turned off
- Interval when both thyristors are conducting is referred to as overlap interval defined by overlap angle γ

Determining the overlap angle γ

- During overlap interval both thyristors considered as short circuits
- Circuit consists of 2 voltage sources V_1 and V_2 and 2 inductances L

$$v_2 - v_1 = 2L \frac{di}{dt}$$

- Difference between phase voltages $v_2 - v_1$ is equal to line voltages V_{21} :

$$v_2 - v_1 = V_{21} = \sqrt{3}V_m \sin(\omega t + \alpha)$$

- $t = 0$ corresponds to instant thyristor T_2 is fired ($\theta = \alpha$)

$$di = \frac{\sqrt{3}V_m}{2L} \sin(\omega t + \alpha) dt$$

- Integrate equation from $t = 0$ to t :

$$i = \int_0^t di = \int_0^t \frac{\sqrt{3}V_m}{2L} \sin(\omega t + \alpha) dt$$

- Substitute $\theta = \omega t + \alpha$

$$\begin{aligned} i &= \frac{\sqrt{3}V_m}{2\omega L} \int_{\alpha}^{\omega t + \alpha} \sin \theta d\theta = \frac{\sqrt{3}V_m}{2\omega L} [-\cos \theta]_{\alpha}^{\omega t + \alpha} \\ &= \frac{\sqrt{3}V_m}{2\omega L} (\cos \alpha - \cos(\omega t + \alpha)) \end{aligned}$$

- Overlap interval is complete when $i = I_L$ and $\omega t = \gamma$

$$I_L = \frac{\sqrt{3}V_m}{2\omega L} [\cos \alpha - \cos(\gamma + \alpha)]$$

or

$$\gamma = \cos^{-1} \left[\cos \alpha - \frac{2\omega L}{\sqrt{3}V_m} I_L \right] - \alpha$$

- Overlap angle depends on
 - Firing angle α
 - Supply frequency ω
 - Supply inductance L
 - Peak phase voltage of the supply V_m
 - Load current I_L

- If $L = 0$:

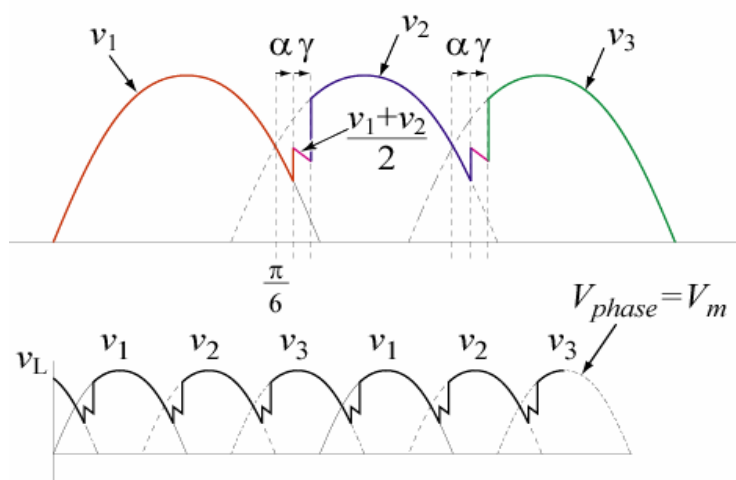
$$\gamma|_{L=0} = -\alpha + \cos^{-1}(\cos \alpha) = -\alpha + \alpha = 0$$

i.e. no overlap

- Same derivation describes the transfer of load current between T_2 and T_3 and between T_3 and T_1

Mean load voltage with overlap

- During overlap both T_1 and T_2 conduct
- Load voltage = mean of phase voltages v_1 and v_2 :



- 3 phase voltages and load voltage for 3-phase half-wave bridge converter with overlap for $\alpha \approx 20^\circ$ and $\gamma \approx 20^\circ$

- $\theta = \pi/6$: Vertical line shows instant at which v_1 and v_2 ; instant at which diode replacing T_2 would conduct \therefore reference for determining α
- $\theta = \pi/6 + \alpha$: T_2 fired; load voltage follows mean of v_1 and v_2
- $\theta = \pi/6 + \alpha + \gamma$: overlap completed; load voltage follows

- Mean load voltage of ideal converter with no overlap:

$$V_{mean} = \frac{3}{2\pi} \int_{\frac{\pi}{6} + \alpha}^{\frac{5\pi}{6} + \alpha} V_m \sin \theta d\theta$$

- $\theta = \pi/6 + \alpha$ until $\theta = \pi/6 + \alpha + \gamma$, load voltage = mean of v_1 and v_2
- $\theta = \pi/6 + \alpha + \gamma$ until $\theta = 5\pi/6 + \alpha$, load voltage = v_2
- And we can express v_1 and v_2 as

$$v_1 = V_m \sin\left(\omega t + \frac{2\pi}{3}\right) \quad v_2 = V_m \sin \omega t$$

- Thus, the mean voltage is,

$$\begin{aligned}
 V_{mean} &= \frac{3}{2\pi} \left\{ \frac{1}{2} \int_{\frac{\pi}{6}+\alpha}^{\frac{\pi}{6}+\alpha+\gamma} V_m \left[\sin\left(\theta + \frac{2\pi}{3}\right) + \sin\theta \right] d\theta + \int_{\frac{\pi}{6}+\alpha+\gamma}^{\frac{5\pi}{6}+\alpha} V_m \sin\theta d\theta \right\} \\
 &= \frac{3V_m}{2\pi} \left\{ \int_{\frac{\pi}{6}+\alpha+\gamma}^{\frac{5\pi}{6}+\alpha} \sin\theta d\theta + \frac{1}{2} \int_{\frac{\pi}{6}+\alpha}^{\frac{\pi}{6}+\alpha+\gamma} \left[\sin\left(\theta + \frac{2\pi}{3}\right) + \sin\theta \right] d\theta \right\} \\
 &= \frac{3V_m}{2\pi} \left\{ \left[-\cos\theta \right]_{\frac{\pi}{6}+\alpha+\gamma}^{\frac{5\pi}{6}+\alpha} + \frac{1}{2} \left[-\cos\left(\theta + \frac{2\pi}{3}\right) \right]_{\frac{\pi}{6}+\alpha}^{\frac{\pi}{6}+\alpha+\gamma} + \frac{1}{2} \left[-\cos\theta \right]_{\frac{\pi}{6}+\alpha}^{\frac{\pi}{6}+\alpha+\gamma} \right\} \\
 &= \frac{3\sqrt{3}V_m}{4\pi} (\cos(\alpha + \gamma) + \cos\alpha)
 \end{aligned}$$

- Finally,

$$V_{mean} = \frac{3\sqrt{3}V_m}{4\pi} [\cos(\alpha + \gamma) + \cos\alpha]$$

- Notice if $L_1 = L_2 = L_3 = 0$, then $\gamma = 0$:

$$V_{mean(ideal)} = \frac{3\sqrt{3}V_m}{2\pi} \cos\alpha$$

as in ideal case

- During overlap intervals, pairs of thyristors are conducting
- This has effect on supply voltage at anodes of thyristors
e.g. voltage on anode of T_1 with overlap

- When conduction changes from T_1 and T_2 or from T_3 to T_1 , v_1 becomes mean of v_1 and v_2 or v_1 and v_3 , respectively, during overlap intervals



Equivalent circuit for the effect of overlap

$$V_{mean} = V_{mean(ideal)} - \Delta V_d$$

where ΔV_d is the change in converter output voltage because of overlap

- Hence we can determine the effect of overlap

$$\begin{aligned} \Delta V_d &= V_{mean(ideal)} - V_{mean} \\ &= \frac{3\sqrt{3}V_m}{2\pi} \cos \alpha - \frac{3\sqrt{3}V_m}{4\pi} [\cos(\alpha + \gamma) + \cos \alpha] \\ &= \frac{3\sqrt{3}V_m}{4\pi} [\cos \alpha - \cos(\alpha + \gamma)] \end{aligned}$$

- α and γ related to load current I_L and other parameters:

$$I_L = \frac{\sqrt{3}V_m}{2\omega L} [\cos \alpha - \cos(\gamma + \alpha)]$$

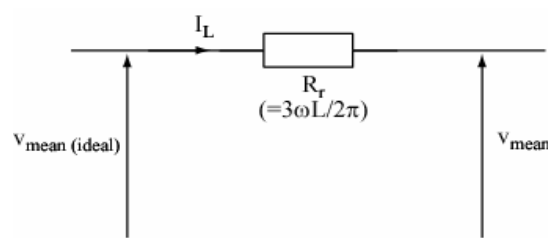
- Effect of overlap:

$$\begin{aligned} \Delta V_d &= \frac{3\sqrt{3}V_m}{4\pi} \frac{2\omega L I_L}{\sqrt{3}V_m} \\ &= \frac{3\omega L}{2\pi} I_L \end{aligned}$$

- Mean load voltage with overlap:

$$\begin{aligned} V_{mean} &= V_{mean(ideal)} - \Delta V_d \\ &= \frac{3\sqrt{3}V_m}{2\pi} \cos \alpha - \frac{3\omega L}{2\pi} I_L \end{aligned}$$

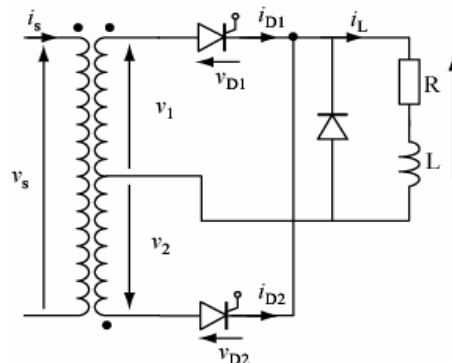
- Reduction in voltage modelled by series resistor:



- R_r models reduction of mean load voltage due to overlap but it does not correctly model power loss because overlap phenomenon is lossless

Overlap example

- Single-phase, full-wave converter with a freewheeling diode is supplied from a 120 V, 50 Hz AC supply with a source inductance of 0.32 mH



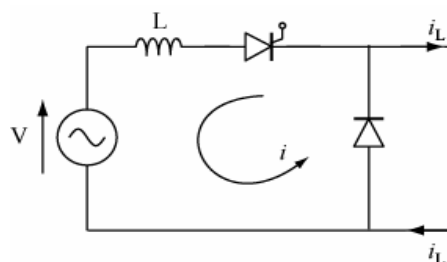
- Assuming that load current is constant at 4 A and firing angle $\alpha = 15^\circ$
- Find overlap angles for:
 - (i) transfer of current from conducting thyristor to diode
 - (ii) transfer of current from commutating diode to a thyristor when the firing angle is 15°

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Solution

- (i) Commutation from thyristor to diode begins when load voltage starts to reverse which we define as the reference instant $t = 0$; assume device forward voltage drops are zero
- Represent transfer of current from initially conducting thyristor to freewheeling diode by means of circulating current i :



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- During overlap, both devices are conducting:

$$v = -V_m \sin \omega t = -L \frac{di}{dt}$$

$$di = \frac{V_m}{L} \sin \omega t dt$$

$$i = \frac{V_m}{L} \int_0^t \sin \omega t dt$$

$$= \frac{V_m}{\omega L} (1 - \cos \omega t)$$

- Commutation complete when $i = I_L$, at which point $\omega t = \gamma_1$; thus

$$I_L = \frac{V_m}{\omega L} (1 - \cos \gamma_1)$$

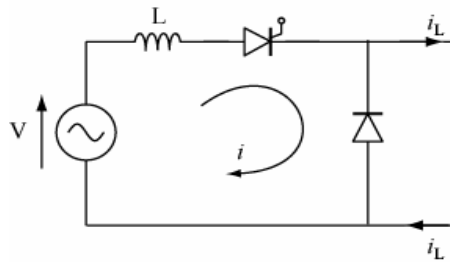
- Hence,

$$\gamma_1 = \cos^{-1} \left[1 - \frac{\omega L I_L}{V_m} \right]$$

- Substitution of values for ω , L , I_L , and V_m , we obtain:

$$\gamma_1 = 3.95^\circ$$

- (ii) Commutation from diode to thyristor begins at instant of firing thyristor; introduce circulating current i to represent change in currents:



- Before thyristor is fired, load current flows through diode
- During overlap interval, both devices are conducting and we can write:

$$v = V_m \sin(\omega t + \alpha) = L \frac{di}{dt}$$

- From which, we have:

$$di = \frac{V_m}{L} \sin(\omega t + \alpha) dt$$

- Integrating from 0 to t :

$$i = \frac{V_m}{\omega L} [\cos \alpha - \cos(\omega t + \alpha)]$$

- Commutation complete when $i = I_L$, at which point $\omega t = \gamma_2$; thus:

$$I_L = \frac{V_m}{\omega L} [\cos \alpha - \cos(\gamma_2 + \alpha)]$$

- Hence:

$$\gamma_2 = \alpha + \cos^{-1} \left[\cos \alpha - \frac{\omega L I_L}{V_m} \right]$$

- Substitution of values for α , ω , L , I_L , and V_m , we obtain:

$$\gamma_2 = 0.516^\circ$$

Summary

- Have shown that finite inductance in AC supply lines in converter causes overlap in which case transfer of load current from one device to another does not take place instantaneously
- Have determined expressions for duration of overlap in converter and its effect on mean load voltage
- Have also modelled effect of overlap on converter output voltage by simple consisting of ideal voltage source and resistor
- Next, we consider operation of fully controlled converters in inverter mode